

Identification of IBR-driven Subsynchronous Oscillations

CIGRE UK C4 Technical Liaison Meeting

Dr Diptargha Chakravorty

Principal Consultant

Siemens Energy

Jan 21, 2025



cigre

For power system expertise

Topics covered

- Project overview
- CIGRE C4 publications
- Frequency domain analysis methods
- Impedance measurement methods
 - Phase to sequence transformation
 - Phase to $\alpha\beta$ transformation
- Test network
 - Method 1 and 2 comparison
 - Sequence Vs $\alpha\beta$ comparison



Project reference and publications

Project Reference

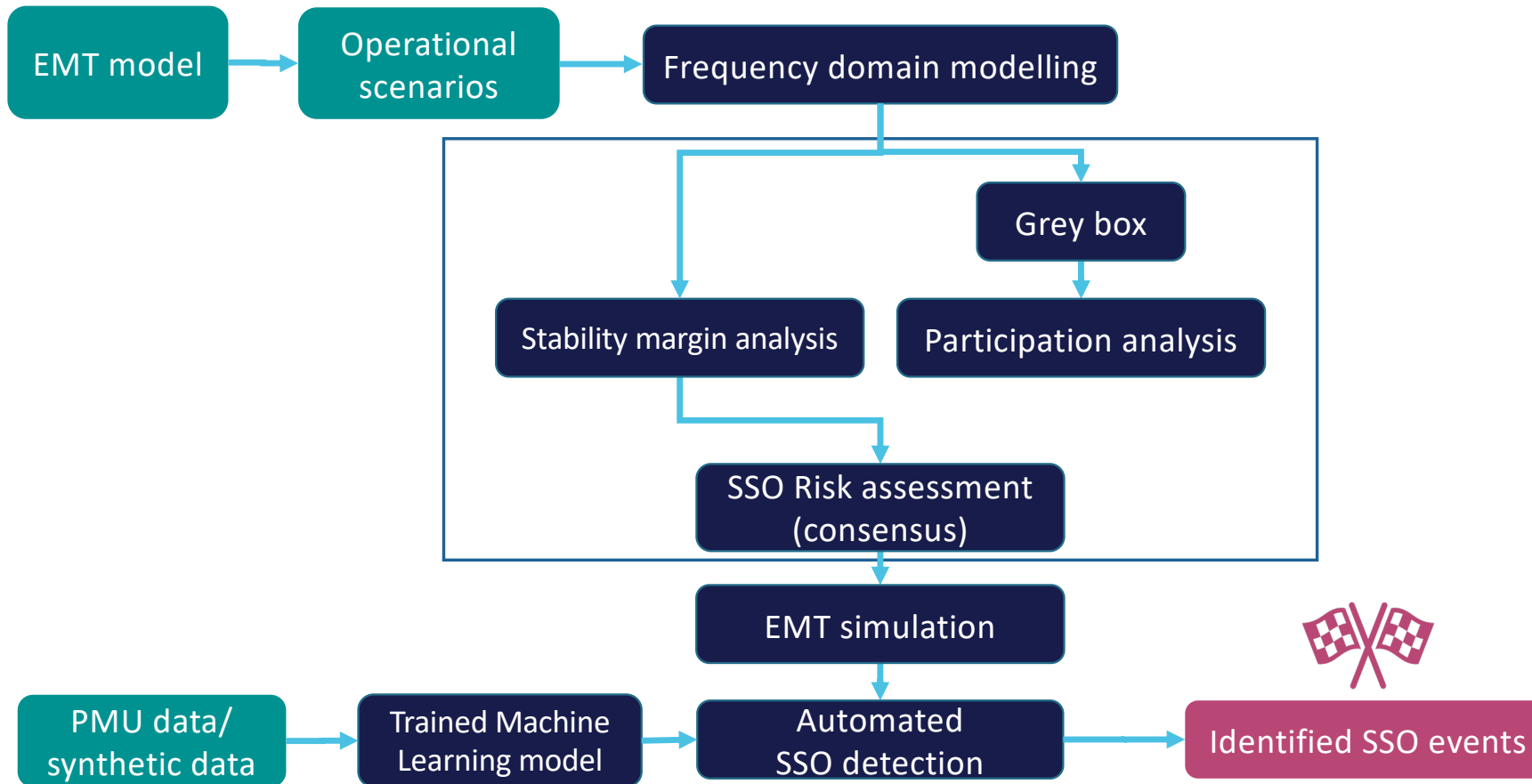


Automated Identification of SSO events

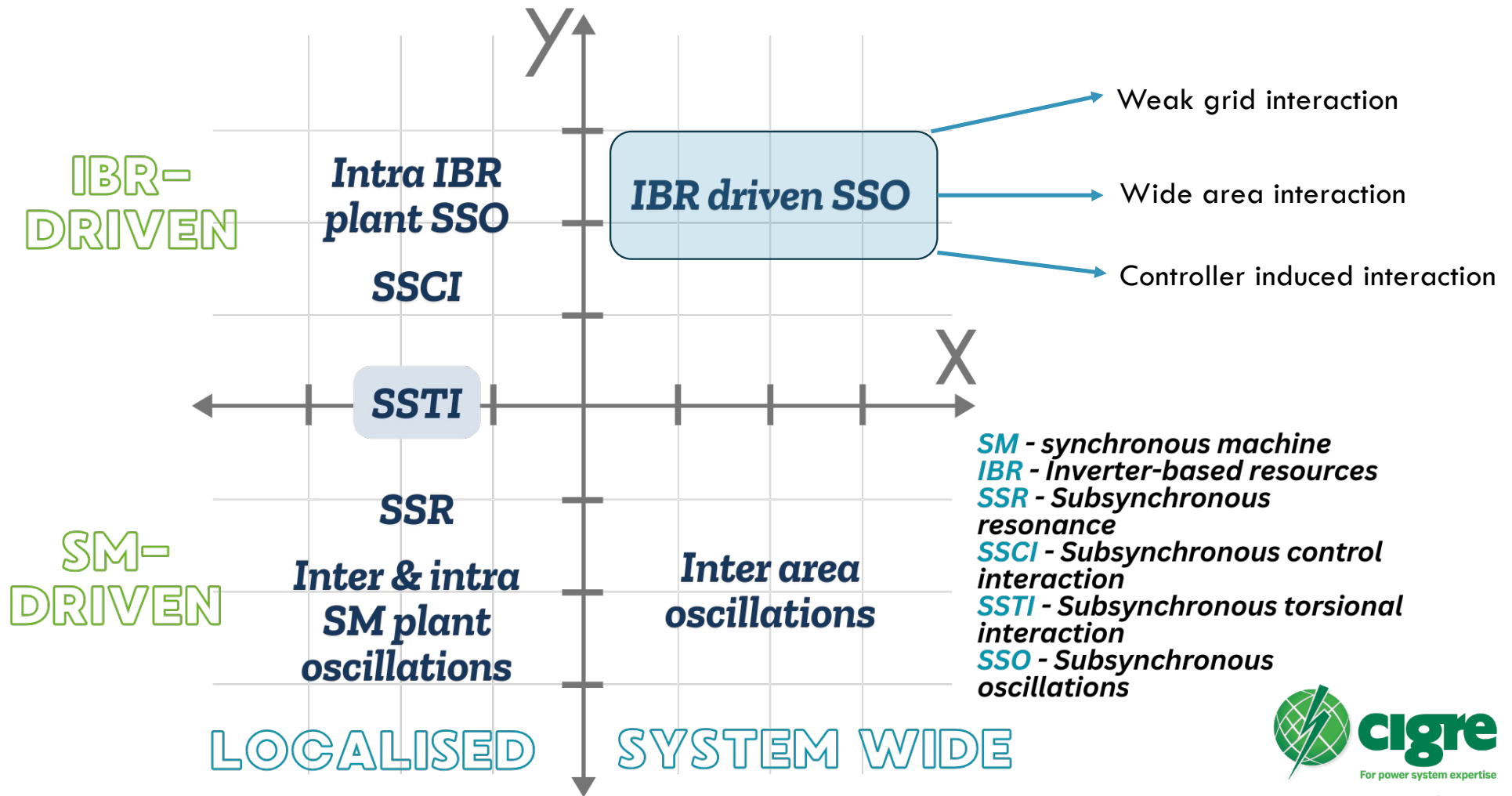
This project explored, developed, and tested a combination of novel frequency domain methodologies and machine learning techniques to identify potential system operating conditions that can lead to Sub-Synchronous Oscillations (SSOs) through an automated control interaction studies framework.



Overall Approach



Types of Subsynchronous Oscillations



CIGRE Publications 1



2024 Paris Session

Paper ID – 11099

C4 Power system technical performance

PS1 Power system dynamic analysis in the energy transition: challenges, opportunities and advances

Framework for Identification of Subsynchronous Oscillations Risks

**Diptargha
CHAKRAVORTY***
TNEI Services Ltd
United Kingdom
diptargha@
tneigroup.com

**Jaime
TRIVINO**
TNEI Services Ltd
United Kingdom
jaime.trivino@
tneigroup.com

**Sami
ABDELRAHMAN**
National Grid ESO
United Kingdom
sami.abdelrahman@
nationalgrideso.com

This paper discusses the developed framework and the associated tool for SSO identification along with an example case study



CIGRE Publications 2



2024 Paris Session

Paper ID – 11096

C4 Power system technical performance

PS1 Power system dynamic analysis in the energy transition: challenges, opportunities and advances

Automatic Detection of Subsynchronous Oscillations

**Diptargha
CHAKRAVORTY***

TNEI Services Ltd

**United Kingdom
diptargha@
tneigroup.com**

**Alexandru
Christian NEAGU
Delft University of
Technology
The Netherlands
a.c.neagu@
tudelft.nl**

**Jochen L.
CREMER
Delft University of
Technology
The Netherlands
j.l.cremer@
tudelft.nl**

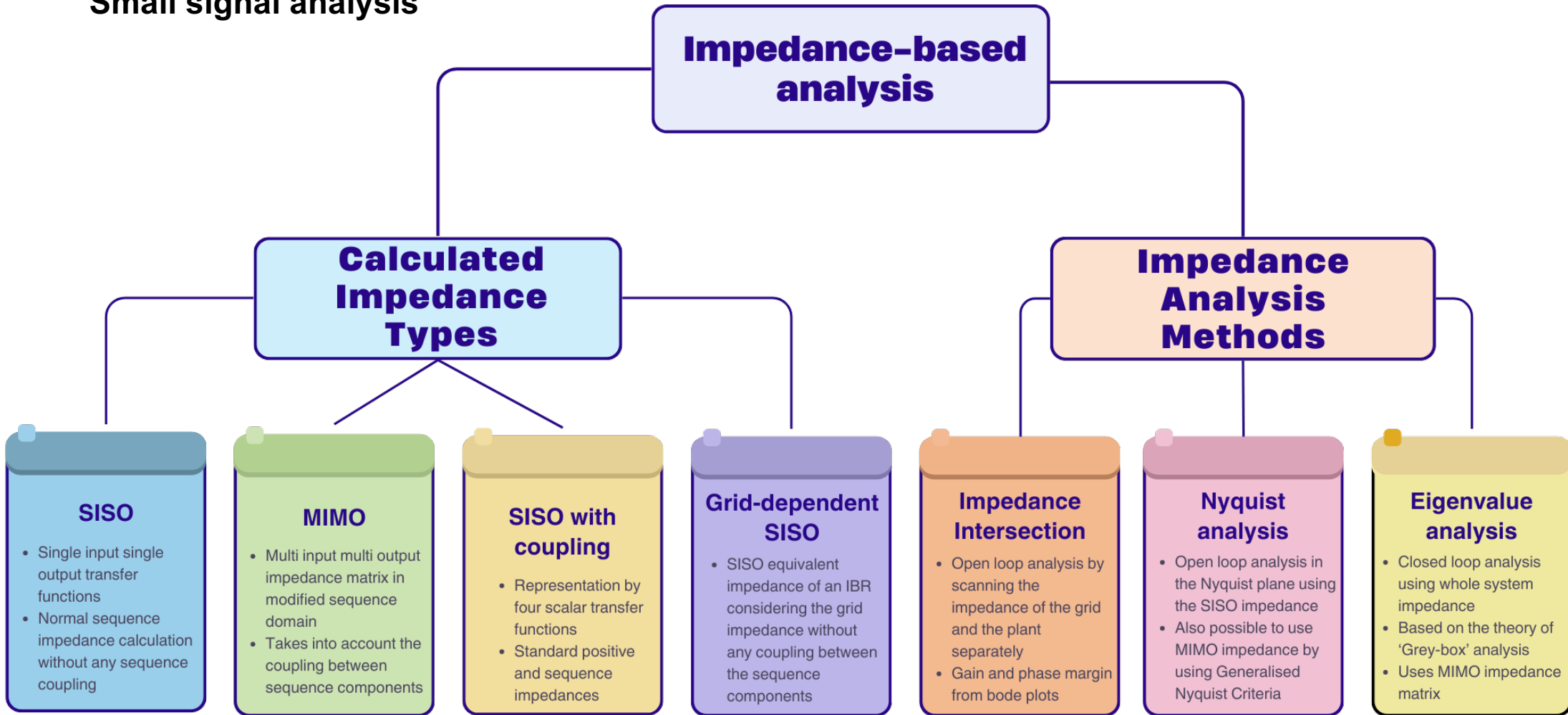
This paper discusses the developed machine learning-based tool for detection of oscillations in measurement data either from PMU or EMT simulations.



Frequency domain analysis methods

Frequency domain analysis

Small signal analysis



SISO Form

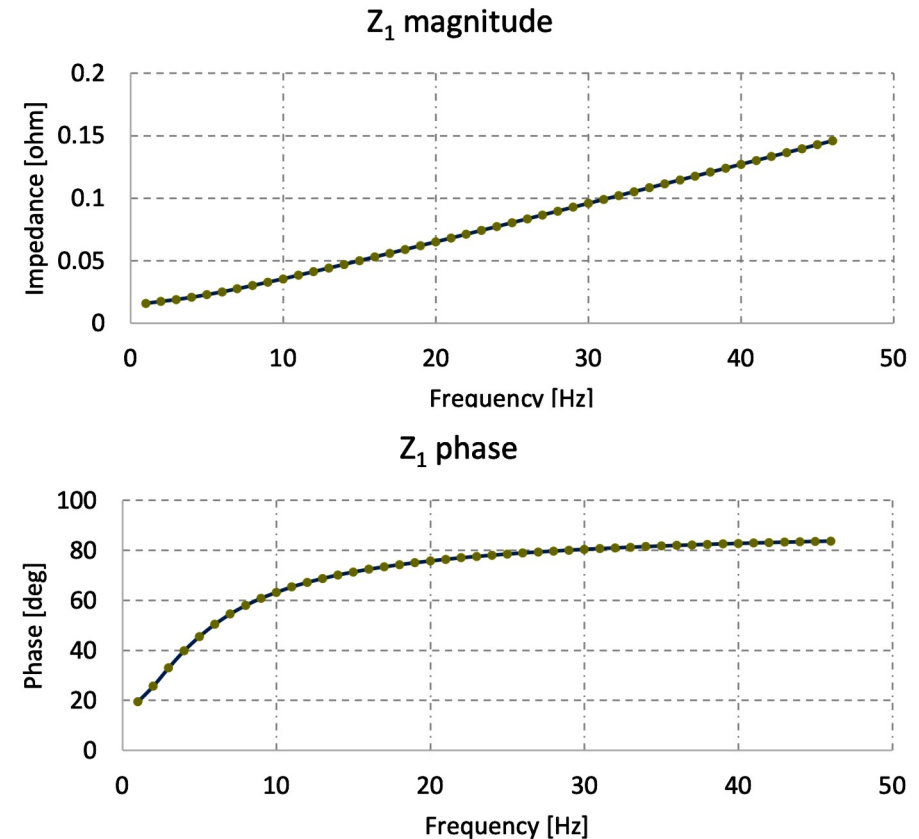
Grid-independent

- If the positive and negative sequence impedances are **decoupled**, then the standard sequence domain impedance can be used.
- Active devices like Inverter-based resources (IBRs) have a frequency-coupled response due to the configuration of controller.
- Mirror Frequency Coupling (MFC) is dominant at low frequencies (**Sub-synchronous range**) and has an impact on stability performance.

Standard
sequence
domain
impedances

$$Y_p = \frac{I_p}{V_p}$$

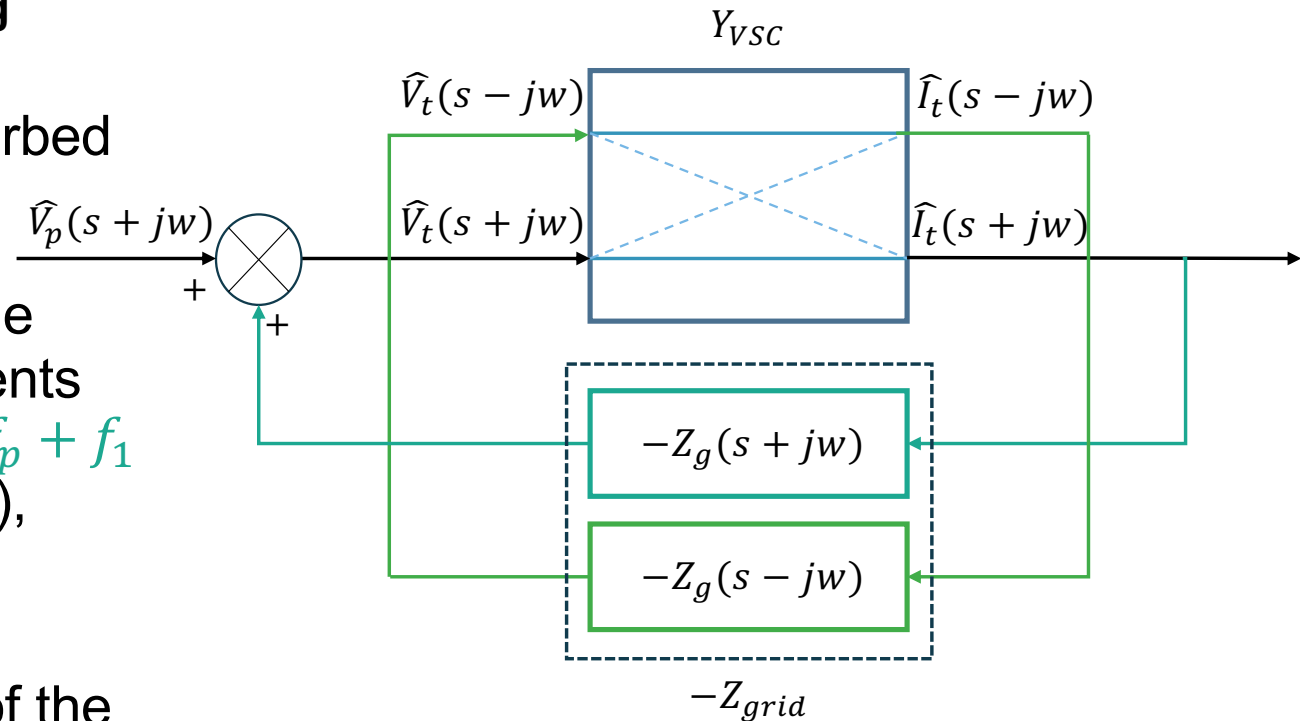
$$Y_n = \frac{I_n}{V_n}$$



MIMO Form

Full matrix with sequence coupling

- When a 3-phase VSC is disturbed from its nominal operation by injecting a perturbation at an arbitrary frequency f_p , then the dominant frequency components in three phase variables are $f_p + f_1$ (+ve seq) and $f_p - f_1$ (-ve seq), where f_1 is the fundamental frequency.
- Voltages at the ac terminals of the



$$v_a = \underbrace{V_1 \cos(2\pi f_1 t + \phi_{v1}) + \hat{V}_p \cos[2\pi(f_p + f_1)t + \phi_{vp}]}_{\text{Positive Sequence}} + \underbrace{\hat{V}_n \cos[2\pi(f_p - f_1)t + \phi_{vn}]}_{\text{Negative Sequence}}$$

S. Shah and L. Parsa, "Impedance Modeling of Three-Phase Voltage Source Converters in DQ, Sequence, and Phasor Domains," in IEEE Transactions on Energy Conversion, vol. 32, no. 3, pp. 1139-1150, Sept. 2017, doi: 10.1109/TEC.2017.2698202.

MIMO Form - continued

Full matrix with sequence coupling

- The small signal behaviour of the 3-phase VSC from its ac terminals can be described as

$$\begin{bmatrix} \hat{I}_p(s + j\omega_1) \\ \hat{I}_n(s - j\omega_1) \end{bmatrix} = \overbrace{\begin{bmatrix} Y_{pp}(s) & Y_{pn}(s) \\ Y_{np}(s) & Y_{nn}(s) \end{bmatrix}}^{Y_{PN}} \begin{bmatrix} \hat{V}_p(s + j\omega_1) \\ \hat{V}_n(s - j\omega_1) \end{bmatrix}$$

- Conventional +ve and -ve seq impedances are related to the diagonal elements of the Y_{PN} as

$$Z_p(s) = \frac{\hat{V}_p(s)}{\hat{I}_p(s)} = \frac{1}{Y_{pp}(s - j\omega_1)}$$

$$Z_n(s) = \frac{\hat{V}_n(s)}{\hat{I}_n(s)} = \frac{1}{Y_{nn}(s + j\omega_1)}$$

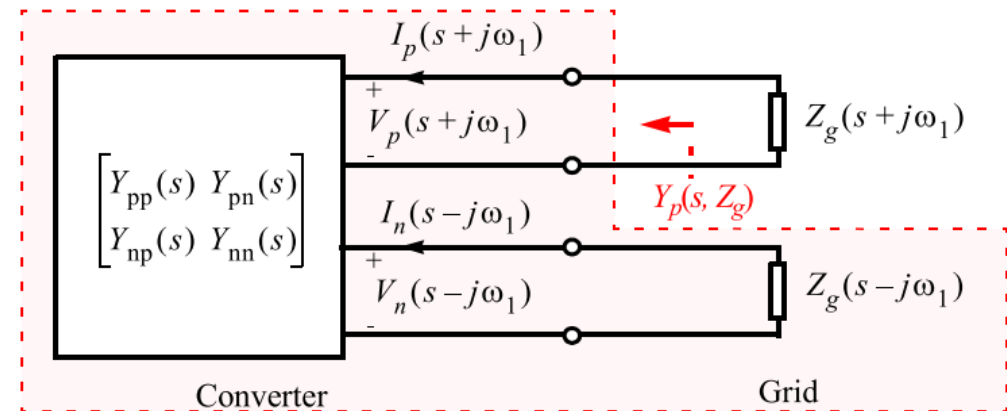
SISO Form

Grid-dependent

- The sequence admittance with frequency coupling of the equipment can be represented using two SISO transfer functions that are dependent on the grid impedance $Z_g(s)$

$$Y_p(s, Z_g) = Y_{pp}(s - j\omega_1) - \frac{Y_{pn}(s - j\omega_1)Y_{np}(s - j\omega_1) \cdot Z_g(s - j2\omega_1)}{1 + Y_{nn}(s - j\omega_1) \cdot Z_g(s - j2\omega_1)}$$

$$Y_n(s, Z_g) = Y_{nn}(s + j\omega_1) - \frac{Y_{pn}(s + j\omega_1)Y_{np}(s + j\omega_1) \cdot Z_g(s + j2\omega_1)}{1 + Y_{nn}(s + j\omega_1) \cdot Z_g(s + j2\omega_1)}$$



- This SISO representation is valid only when the grid (equivalent source) does not exhibit sequence coupling in its sequence impedance.
- This representation is not suitable for aggregating the sequence admittance of different components in a network.

SISO Form

4 scalar transfer functions

- The sequence admittance with frequency coupling can also be represented using four scalar transfer functions.
- Here Y_p and Y_n are standard positive and negative sequence admittances
- And Y_{cp} and Y_{cn} are the admittances at coupling frequencies

$$\left. \begin{aligned} Y_p(s) &= \frac{I_p(s)}{V_p(s)} \\ Y_{cp}(s) &= \frac{I_n(s-j2\omega_1)}{V_p(s)} \end{aligned} \right\} \text{where } V_n(s - j2\omega_1) = 0$$

$$\left. \begin{aligned} Y_n(s) &= \frac{I_n(s)}{V_n(s)} \\ Y_{cn}(s) &= \frac{I_p(s+j2\omega_1)}{V_n(s)} \end{aligned} \right\} \text{where } V_p(s + j2\omega_1) = 0$$

- These transfer functions are related to the elements of the sequence admittance matrix by

$$\begin{aligned} Y_p(s) &= Y_{pp}(s - j\omega) \\ Y_{cp}(s) &= Y_{np}(s - j\omega) \end{aligned}$$

$$\begin{aligned} Y_n(s) &= Y_{nn}(s + j\omega) \\ Y_{cn}(s) &= Y_{pn}(s + j\omega) \end{aligned}$$

Impedance measurement approaches

MIMO Form - measurement approaches

$$\begin{bmatrix} \hat{i}_p(s + j\omega_1) \\ \hat{i}_n(s - j\omega_1) \end{bmatrix} = \overbrace{\begin{bmatrix} Y_{pp}(s) & Y_{pn}(s) \\ Y_{np}(s) & Y_{nn}(s) \end{bmatrix}}^{Y_{PN}} \begin{bmatrix} \hat{v}_p(s + j\omega_1) \\ \hat{v}_n(s - j\omega_1) \end{bmatrix}$$

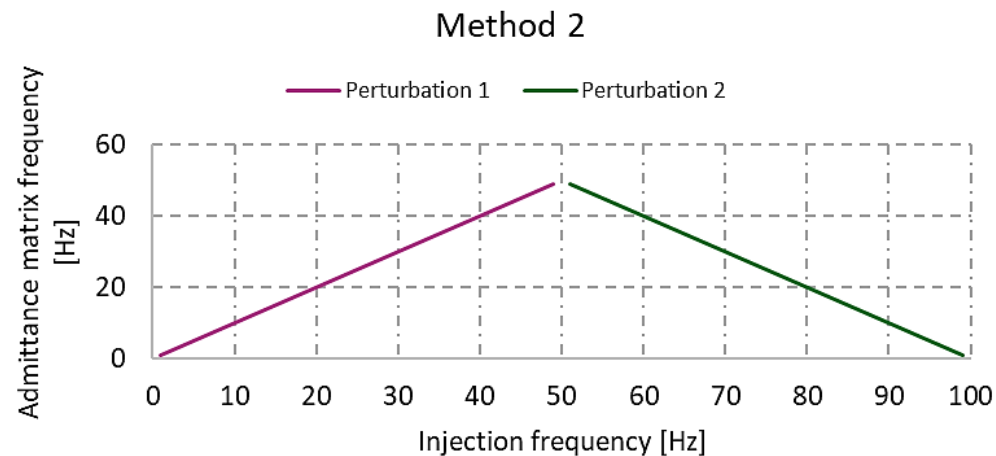
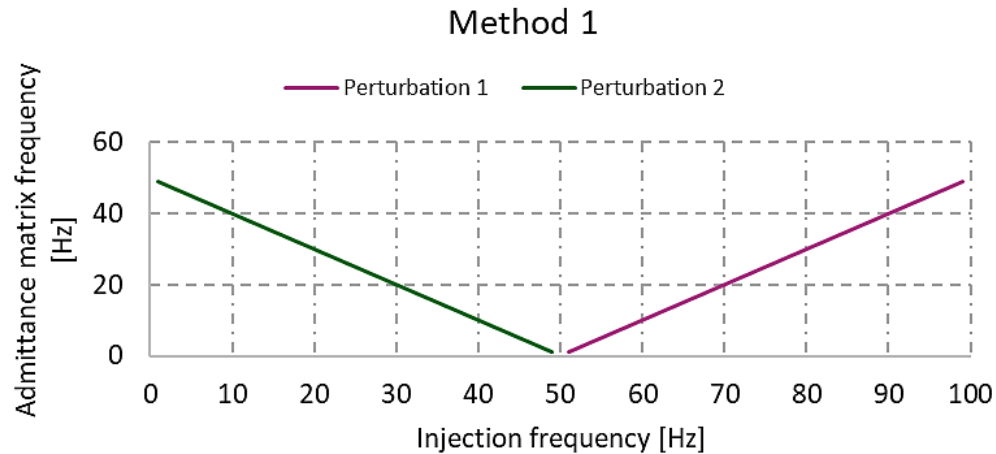
Method 1

$$\begin{bmatrix} \hat{i}_{\alpha\beta}(s) \\ \hat{i}_{\alpha\beta}^*(s - 2j\omega) \end{bmatrix} = \overbrace{\begin{bmatrix} Y_{\alpha\beta 11}(s) & Y_{\alpha\beta 12}(s) \\ Y_{\alpha\beta 21}(s) & Y_{\alpha\beta 22}(s) \end{bmatrix}}^{Y_{\alpha\beta}} \begin{bmatrix} \hat{v}_{\alpha\beta}(s) \\ \hat{v}_{\alpha\beta}^*(s - 2j\omega) \end{bmatrix}$$

Method 2

MIMO Form – Methods comparison

Injection frequency



- No major differences except **Method 1** starts from the fundamental and approaches the ends while **Method 2** starts from the ends and approaches the fundamental.
- Also, **Perturbation 1** measures the positive sequence impedance between the **50-100Hz range** in **Method 1** while the same perturbation measures the impedance between **1-50Hz range** in **Method 2**.

MIMO Form – Sequence domain

Self and mutual impedance terms

Symmetrical, coupled, linear, three-phase system

$$\begin{bmatrix} v_a \\ v_b \\ v_c \end{bmatrix} = \begin{bmatrix} L_S & M & M \\ M & L_S & M \\ M & M & L_S \end{bmatrix} \frac{d}{dt} \begin{bmatrix} i_a \\ i_b \\ i_c \end{bmatrix}$$

Conversion to sequence domain

$$A^{-1} \mathbf{A} \begin{bmatrix} v_0 \\ v_1 \\ v_2 \end{bmatrix} = A^{-1} \begin{bmatrix} L_S & M & M \\ M & L_S & M \\ M & M & L_S \end{bmatrix} A \frac{d}{dt} \begin{bmatrix} i_0 \\ i_1 \\ i_2 \end{bmatrix}$$

$$\begin{bmatrix} v_0 \\ v_1 \\ v_2 \end{bmatrix} = \begin{bmatrix} L_S + 2M & 0 & 0 \\ 0 & L_S - M & 0 \\ 0 & 0 & L_S - M \end{bmatrix} \frac{d}{dt} \begin{bmatrix} i_0 \\ i_1 \\ i_2 \end{bmatrix}$$

$$\mathbf{A} = \begin{bmatrix} 1 & 1 & 1 \\ 1 & a^2 & a \\ 1 & a & a^2 \end{bmatrix}$$

where $a = -0.5 + j0.86$



MIMO Form – Sequence domain

Self and mutual impedance terms

$$\begin{bmatrix} v_0 \\ v_1 \\ v_2 \end{bmatrix} = \begin{bmatrix} L_S + 2M & 0 & \overset{Z_{13}}{0} \\ 0 & L_S - M & 0 \\ \underset{Z_{31}}{0} & 0 & L_S - M \end{bmatrix} \frac{d}{dt} \begin{bmatrix} i_0 \\ i_1 \\ i_2 \end{bmatrix}$$

If the mutual couplings across phases are not equal (asymmetrical three phase system), then the off-diagonal terms will be non-zero and

$$Z_{12} = Z_{21}^* \quad Z_{13} = Z_{31}^* \quad Z_{23} = Z_{32}^* \quad \& \quad Z_{22} = Z_{33}$$

MIMO Form – $\alpha\beta$ frame

Self and mutual impedance terms

Space phasor representation of three phase quantities

$$\vec{f}(t) = \frac{2}{3} \left[e^{j0} f_a(t) + e^{j\frac{2\pi}{3}} f_b(t) + e^{j\frac{4\pi}{3}} f_c(t) \right]$$

Space phasor projection on cartesian coordinates

$$\vec{f}(t) = f_\alpha(t) + jf_\beta(t)$$

Phase quantities to $\alpha\beta$ quantities

$$\begin{bmatrix} f_\alpha(t) \\ f_\beta(t) \end{bmatrix} = \frac{2}{3} \mathbf{C} \begin{bmatrix} f_a(t) \\ f_b(t) \\ f_c(t) \end{bmatrix}$$

$$\mathbf{C} = \begin{bmatrix} 1 & -\frac{1}{2} & -\frac{1}{2} \\ 0 & \frac{\sqrt{3}}{2} & -\frac{\sqrt{3}}{2} \end{bmatrix}$$

MIMO Form – $\alpha\beta$ frame

Self and mutual impedance terms

Symmetrical, coupled, linear, three-phase system

$$\begin{bmatrix} v_a \\ v_b \\ v_c \end{bmatrix} = \begin{bmatrix} L_S & M & M \\ M & L_S & M \\ M & M & L_S \end{bmatrix} \frac{d}{dt} \begin{bmatrix} i_a \\ i_b \\ i_c \end{bmatrix}$$

Phase quantities to $\alpha\beta$ quantities

$$\frac{2}{3} CC^T \begin{bmatrix} v_\alpha \\ v_\beta \end{bmatrix} = \frac{2}{3} C \begin{bmatrix} L_S & M & M \\ M & L_S & M \\ M & M & L_S \end{bmatrix} C^T \frac{d}{dt} \begin{bmatrix} i_\alpha \\ i_\beta \end{bmatrix}$$

Self and mutual inductance in $\alpha\beta$ frame

$$\begin{bmatrix} v_\alpha \\ v_\beta \end{bmatrix} = \begin{bmatrix} L_S - M & \alpha\beta_{12} \\ \alpha\beta_{21} & L_S - M \end{bmatrix} \frac{d}{dt} \begin{bmatrix} i_\alpha \\ i_\beta \end{bmatrix}$$

If the mutual couplings across phases are not equal (asymmetrical three phase system), then the off-diagonal terms will be non-zero and

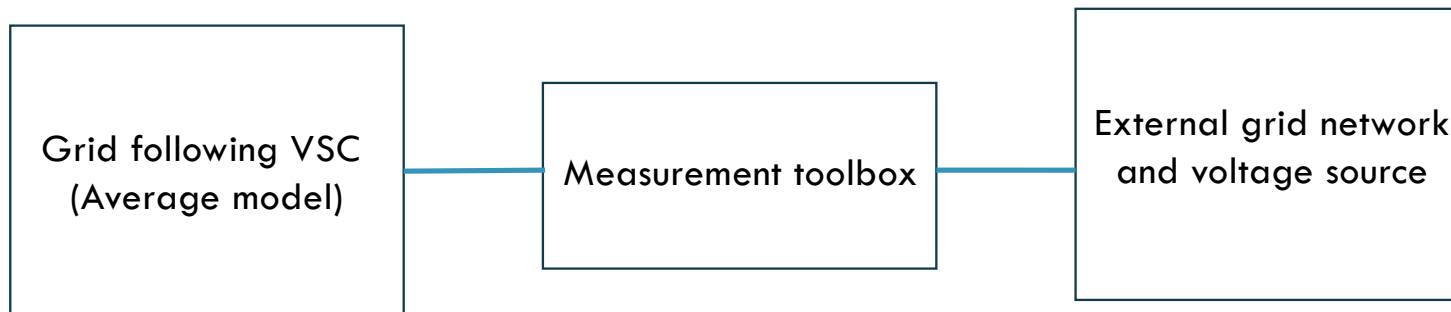
$$\alpha\beta_{11} \neq \alpha\beta_{22}, \alpha\beta_{12} = \alpha\beta_{21}$$

Measurement methods comparison

MIMO Form – Comparison

Test network impedance measurement

Modelled in PSCAD



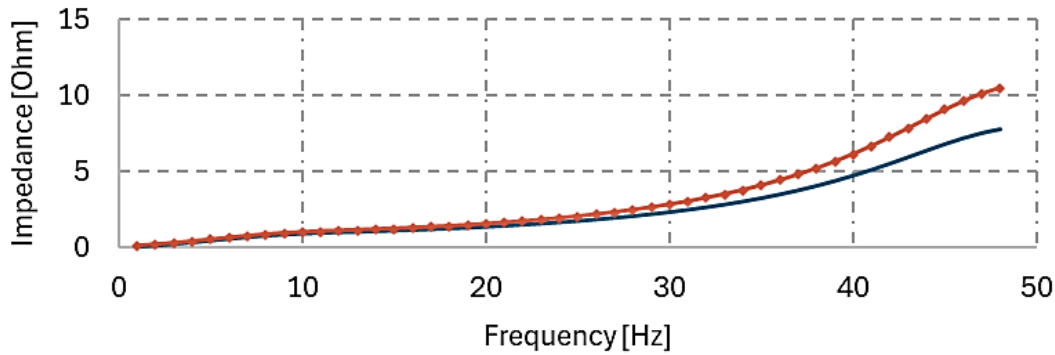
- 50 Hz system, VSC in PQ control mode
- Impedance measured in the subsynchronous range
- Method 1 and Method 2 used to estimate the impedance of the converter
- Sequence impedance compared to $\alpha\beta$ impedance

Modified Seq impedance – Zpp & Znn

Method 1 & Method 2

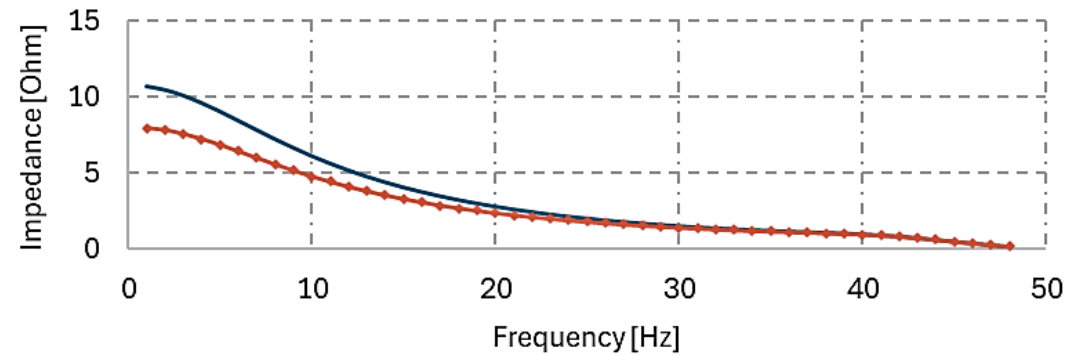
Method 1

— Zpp — Znn



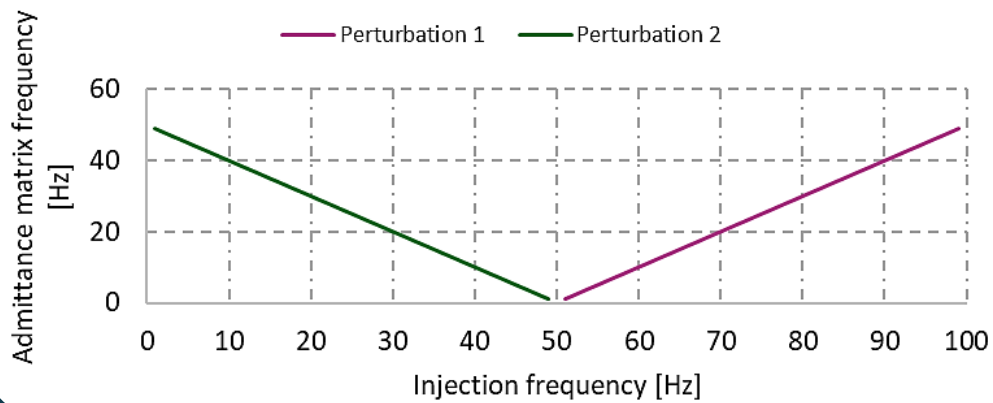
Method 2

— Zpp — Znn



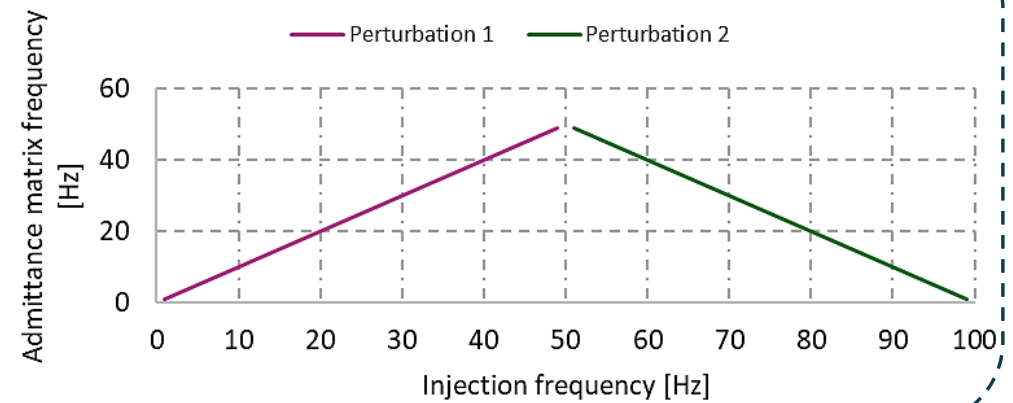
Method 1 $(s + jw)$ & $(s - jw)$

— Perturbation 1 — Perturbation 2



Method 2 (s) & $(s - 2jw)$

— Perturbation 1 — Perturbation 2

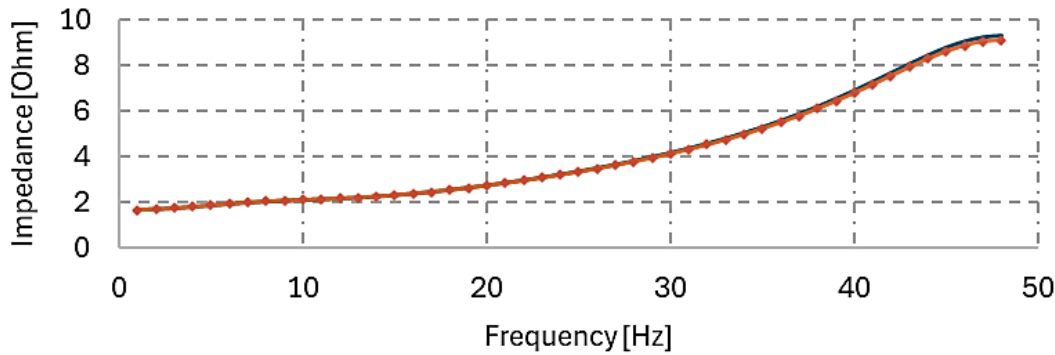


Modified Seq impedance – Zpn & Znp

Method 1 & Method 2

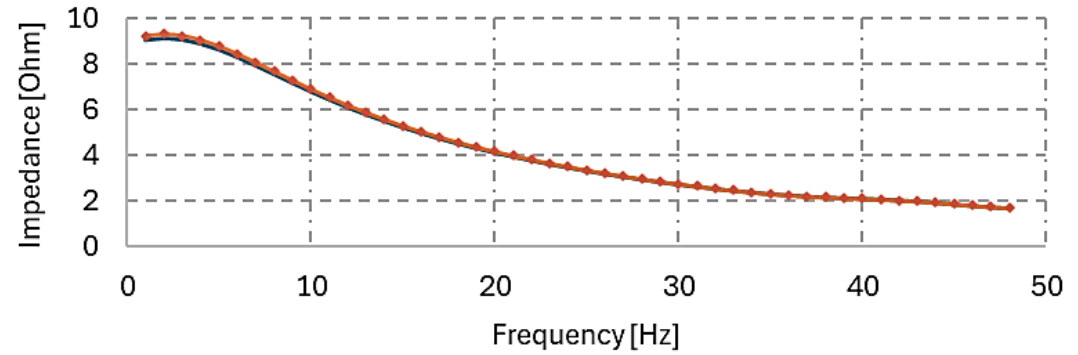
Method 1

— Zpn — Znp



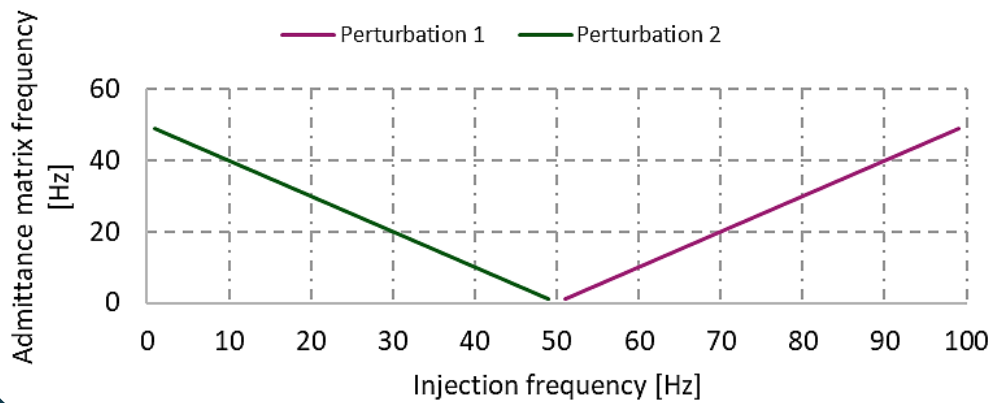
Method 2

— Zpn — Znp



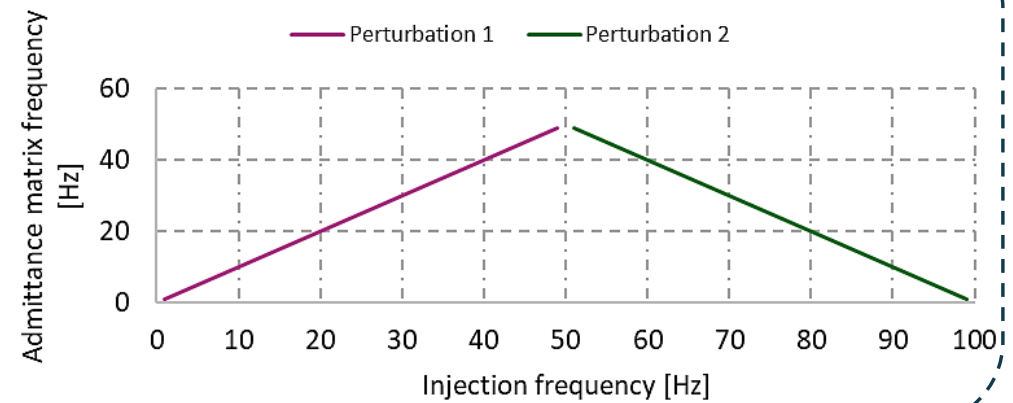
Method 1 $(s + jw)$ & $(s - jw)$

— Perturbation 1 — Perturbation 2



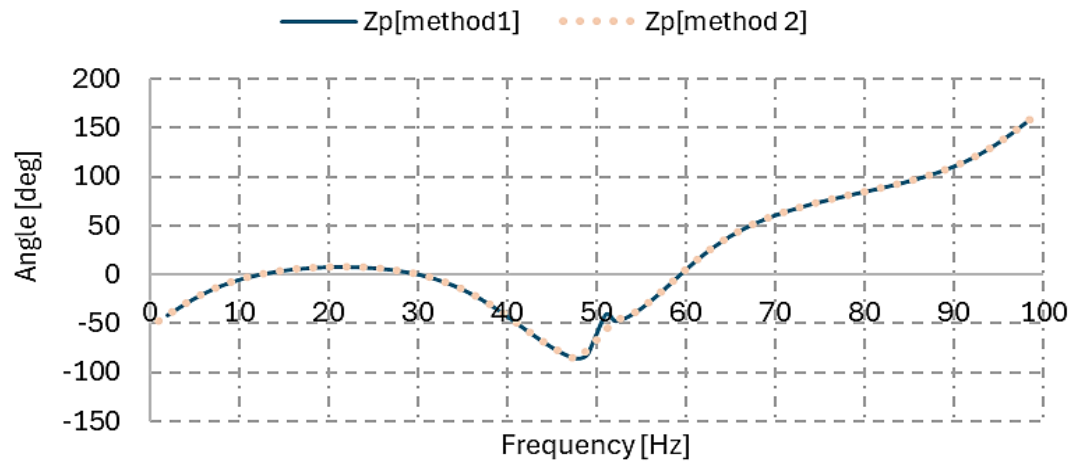
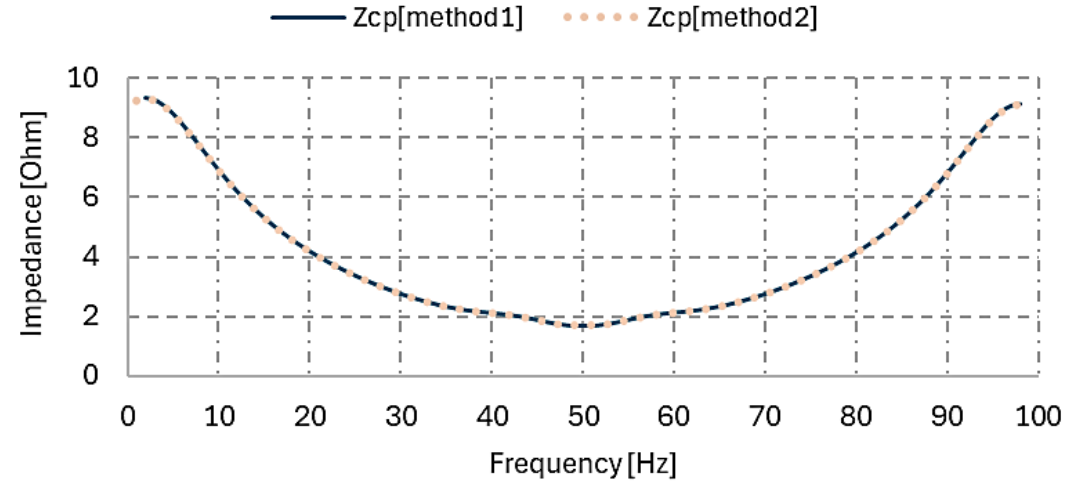
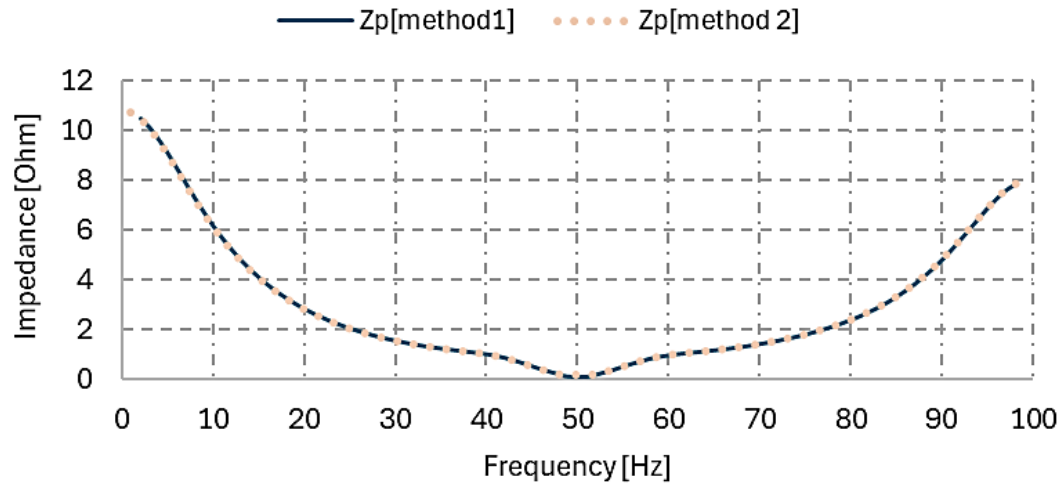
Method 2 (s) & $(s - 2jw)$

— Perturbation 1 — Perturbation 2



Seq impedance comparison

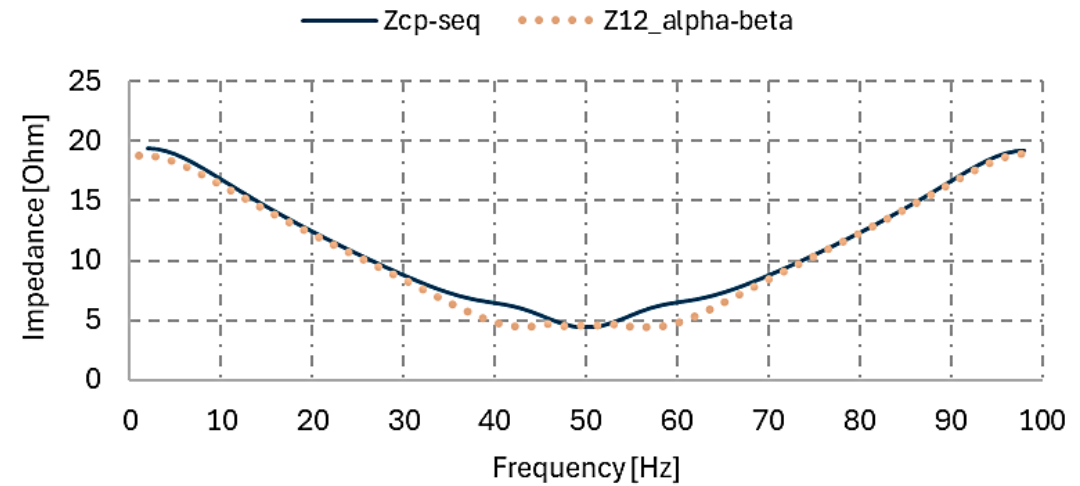
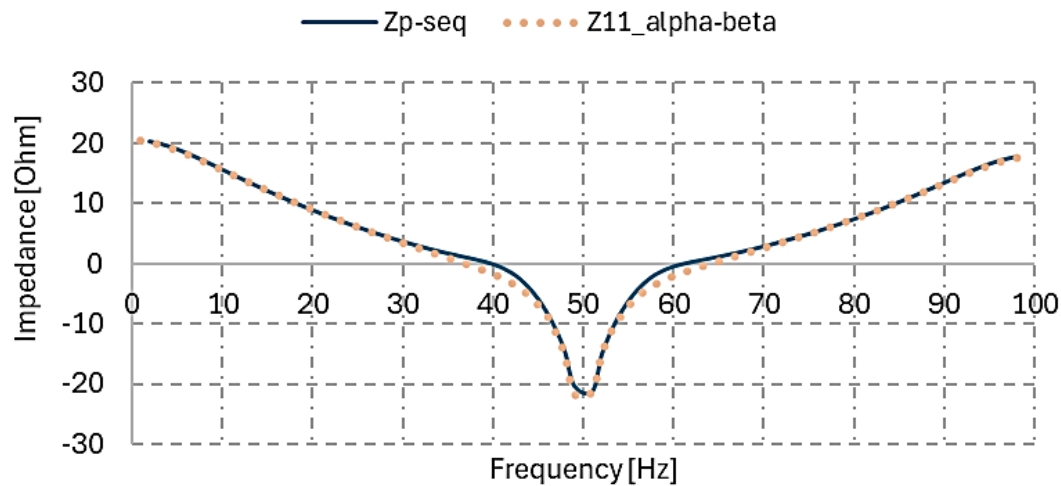
Method 1 & Method 2



$$Z_{p-method2}(s) = Z_{p-method1}^*(s)$$
$$Z_{cp-method2}(s) = Z_{cp-method1}^*(s)$$

MIMO Form – Comparison

Sequence vs $\alpha\beta$ impedance



- Impedances calculated in the $\alpha\beta$ frame corresponds to the sequence impedance calculated by Method 1
- There could be differences between the two domain if there are unequal mutual inductances between the phases
- Finite coupling impedance observed in both methods, which is expected for a converter system

Thank You for listening!